

Electronics – A/D and D/A converters

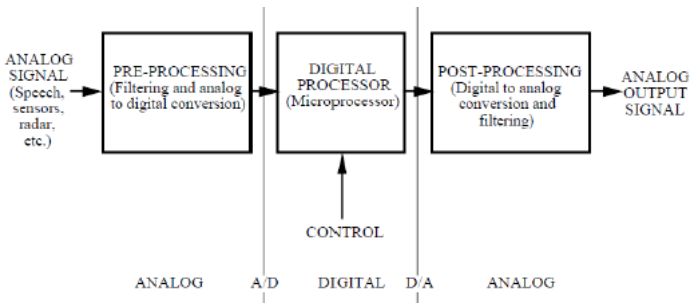
Prof. Márta Rencz, Gábor Takács, Dr. György Bognár,
Dr. Péter G. Szabó

BME DED

December 1, 2014

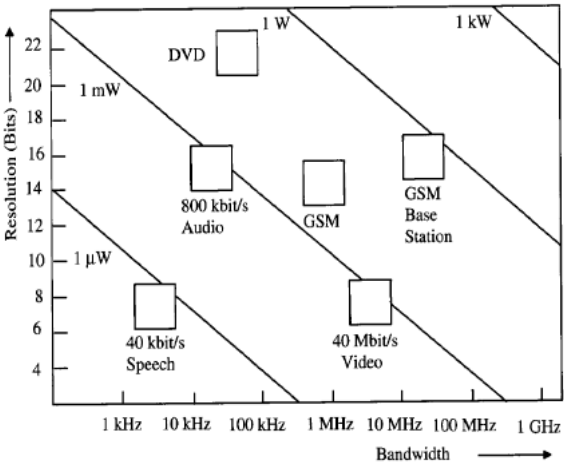
Introduction

- The **world is analog, signal processing nowadays is digital.**
- The transition between the two domains is done using analog-to-digital (A/D) and digital-to-analog (D/A) converters:
 - 1 the input signal is first processed (amplified and filtered),
 - 2 converted to a digital form (A/D conversion),
 - 3 the digital signal is processed
 - 4 and converted back to analog at the output (D/A conversion).

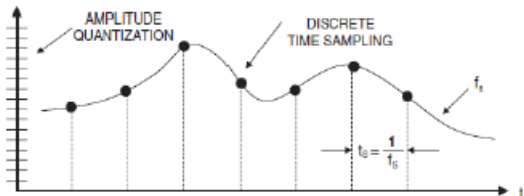


Resolution, bandwidth and energy

The higher the bandwidth or the resolution of a signal, the more energy it takes to convert it.



Sampling



In the course of the A/D conversion of an analog signal, samples are taken at a T_s interval.

The proximity of the digital function to the original analog one is a function of the **sampling frequency**:

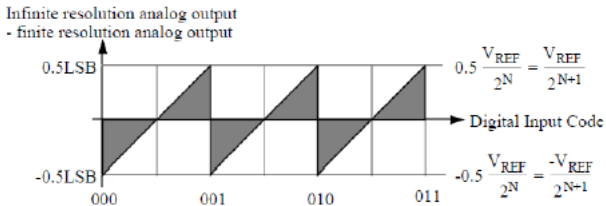
$$f_s = \frac{1}{T_s}$$

Nyquist-Shannon sampling theorem

If **highest frequency in the spectrum** of the input signal is f_{max} then it is **completely determined** by sampling its values at:

$$f_s \geq 2 \cdot f_{max}$$

Quantization error



- Digital sampling introduces **quantization error**. It manifests as a low-level noise added to the reconstructed signal.

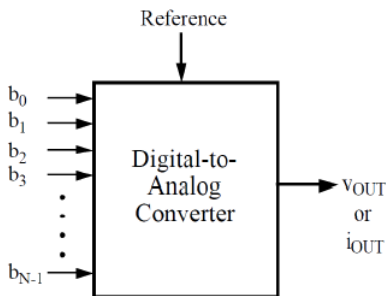
Signal-to-noise ratio (SNR)

$$SNR(\text{dB}) = 1.76 + 6.02 \cdot N \text{ dB} \approx 6N \text{ dB}$$

- E.g. the theoretical SNR of a CD recording (16 bit):

$$SNR_{CD} > 96 \text{ dB}$$

D/A conversion



$$V_{out} = \frac{V_{ref}}{2^N} \cdot B = V_{LSB} \cdot B$$

where

- V_{ref} is the reference voltage,
- N is the resolution of the conversion,
- B is the binary value,
- V_{LSB} is the voltage that corresponds to the LSB value.

The ideal D/A converter

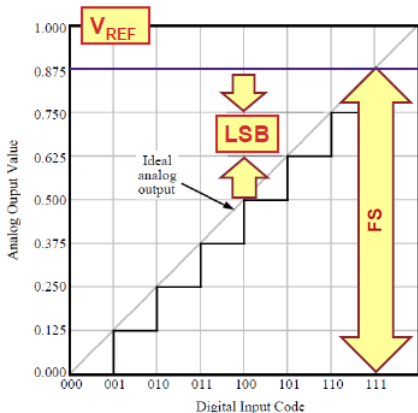
Full scale (FS)

$$V_{out,max} = \frac{V_{ref}}{2^N} (2^N - 1) = FS$$

$$V_{out,min} = 0$$

The LSB voltage

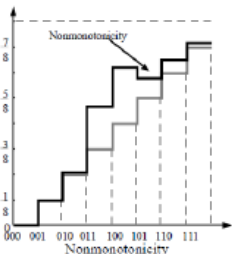
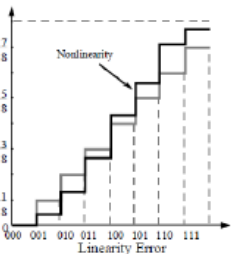
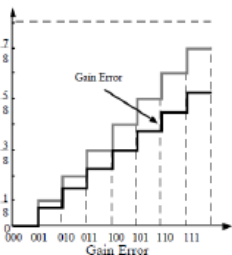
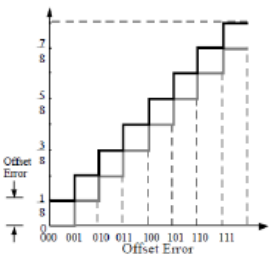
$$V_{LSB} = \frac{V_{ref}}{2^N}$$



The properties of a non-ideal D/A converter

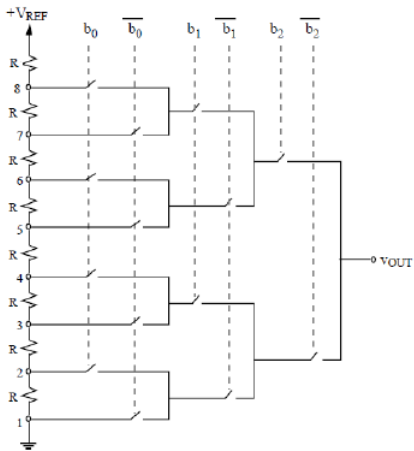
Errors of D/A converters:

- offset error,
- gain error,
- nonlinearity error,
- monotonicity error.

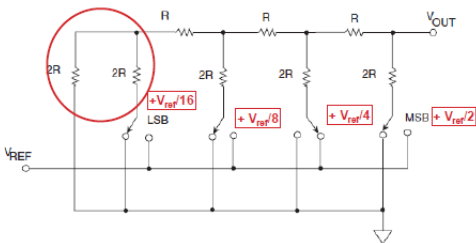


Parallel (direct) D/A conversion

- The reference voltage is divided into 2^N parts.
- The **bits of the binary value control switches** that connect the right analog value to the output.
- This is an **analog multiplexer**.
- An analog switch can be realized using a **CMOS transfer gate**.
- It requires **identical resistors**.
- It is monotonic per construction.
- For N bits 2^N resistors a needed.

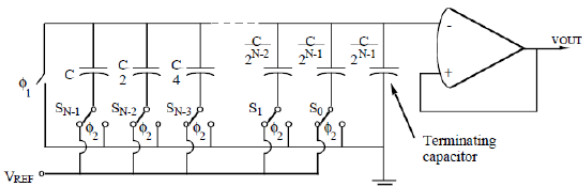


R-2R D/A converter



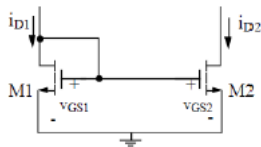
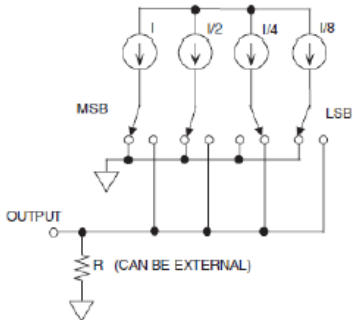
- It can be proven using the theorem of superposition that the **voltage connected to the output when a switch is on corresponds to the binary weight.**
- The **advantage** of this solution is that although accurate resistors are hard to realize in ICs, accurate resistance ratios can be very accurate.
- It contains resistors of value R merely ($2R$ is realized with two R s).
- For N bits $3N + 1$ resistors are needed.

Weighted capacitor D/A converter



- In φ_1 phase **every capacitor is discharged.**
- In the φ_2 phase, if the input is
 - logic 1, the **reference voltage,**
 - logic 0, **ground potential**
 is **connected to the corresponding capacitor.**
- The capacitance of capacitors connected in parallel adds up.

Current switched D/A converter

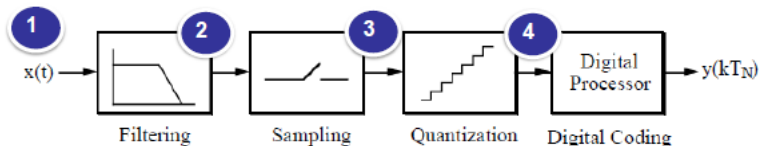


If the transistors are identical:

$$I_{D1} = I_{D2}$$

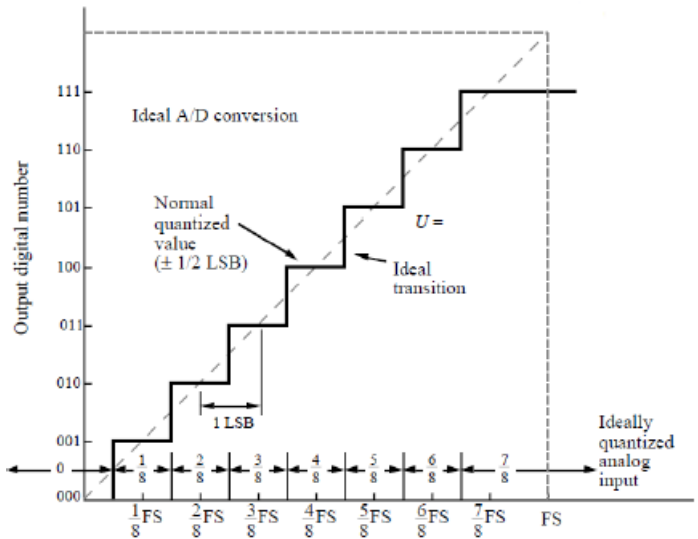
The **currents are switched** using current mirrors connected in parallel **according to the binary weight**.

The process of A/D conversion



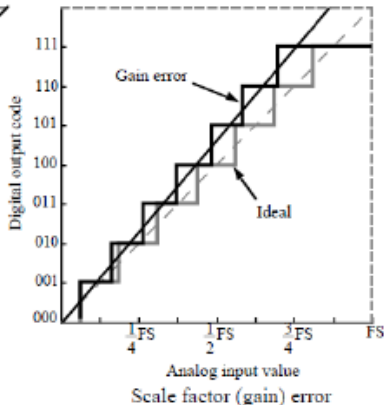
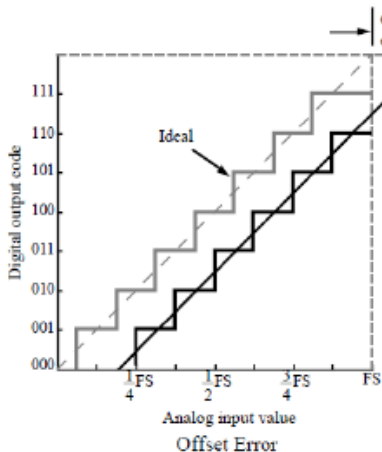
- 1 Anti aliasing filter:** a low-pass filter used to filter out components above f_{max}
- 2 Sampling**
- 3 Quantization**
- 4 Digital encoding**

The ideal A/D converter



LSB: is the voltage corresponding to least significant bit.

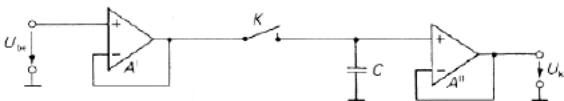
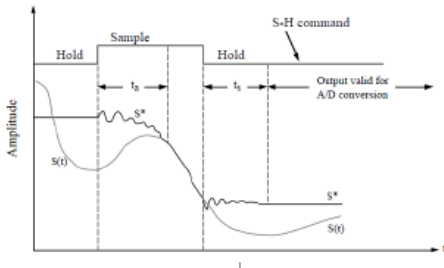
Errors of non-ideal A/D converters



The error types are similar to those of D/A converters.

The sample and hold (S/H) circuit

- When **switched on**, the output copies the input voltage.
- When **switched off**, the last input value is held while an A/D conversion is performed.

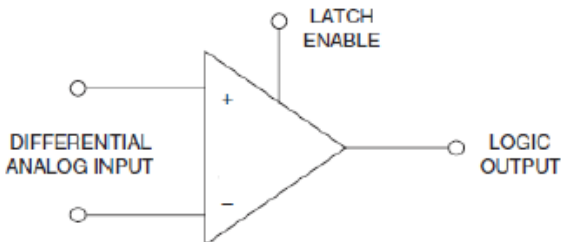


The value is held in the capacitor:

- by the time the switch is turned off, the **capacitor is charged to V_{in}** ,
- a **voltage follower at the output ensures that the voltage of the capacitor is constant** during the conversion.

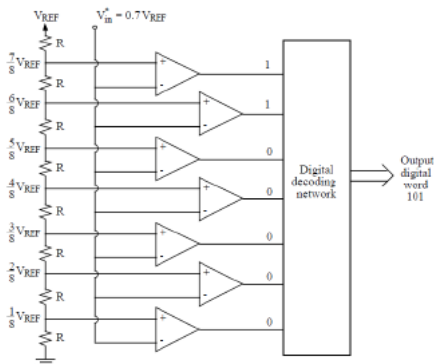
Comparator

- A **comparator**'s output is
 - logic 1, if $V_+ > V_-$,
 - logic 0, if $V_+ < V_-$.
- It's symbol is the same as the operational amplifier's, but they are not the same.

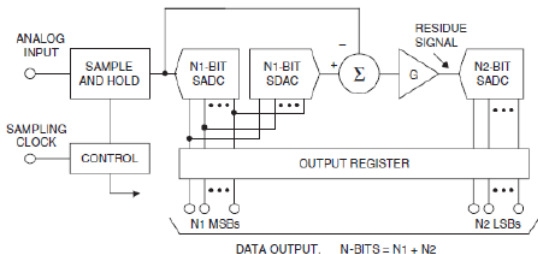


Flash A/D converter

- The reference voltage is divided into 2^N parts.
- Comparators are used to compare each value in the divider with the input.
- The output of the comparators is a **thermometric code**:
 - the bits below the input value are logic 0,
 - the bits above it are logic 1.
- This code needs to be converted to binary.
- For a resolution of N bits 2^N resistors are needed, thus these converters need a very large chip area – they are fabricated with a resolution of 8 – 9 bits at most.



Cascaded flash A/D converter

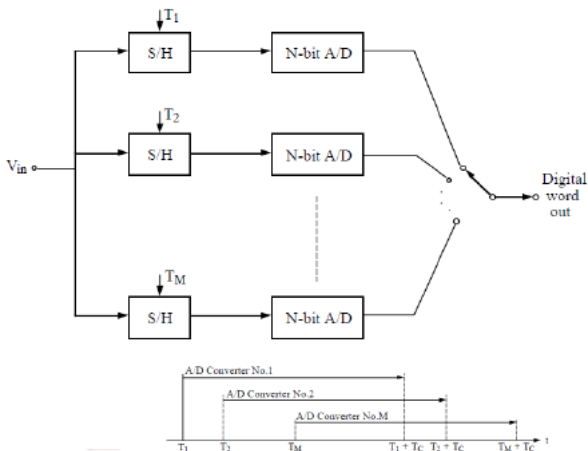


- 1 the high bits are converted,
- 2 this value is subtracted from the input,
- 3 the rest is converted using the other converter.

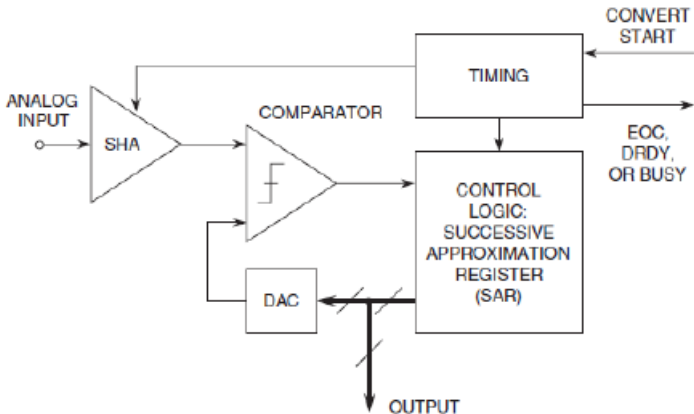
- The resolution is $N = N_1 + N_2$ bits.
- The length of the conversion:
 $t_{A/D} + t_{D/A} + t_{subtraction} + t_{A/D}$
- $2^{N_1} + 2^{N_2} - 2$ converters needed instead of $2^{N_1+N_2} - 1$
- This is a **trade-off between speed and chip area.**

High-speed A/D conversion

- M slow converters work in turns.
- The overall sampling frequency can be increased M times.



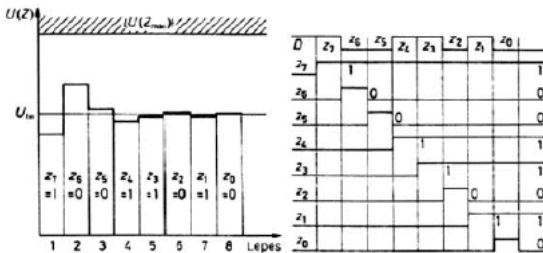
Successive approximation D/A conversion I.



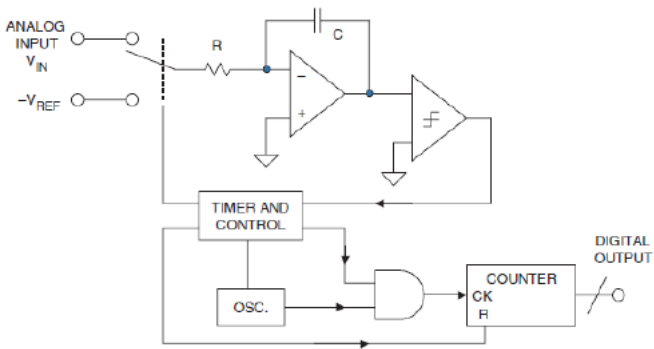
N bits are calculated in N steps.

Successive approximation D/A conversion II.

- **At the beginning of the conversion** the MSB bit is 1, the rest is 0.
- The **input value** is compared to the **binary value** converted to analog by the D/A converter. \hat{u}
- **If the DAC's output is bigger**, the bit is set to zero, the one below it is set to 1.
- This is **done for every bit**.
- The length of the conversion: $N \cdot T_{step}$.

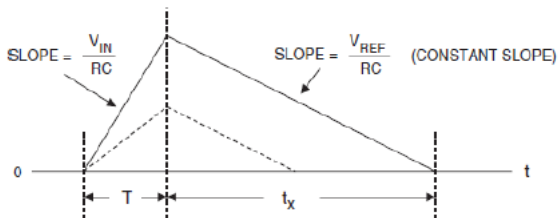


Dual-slope A/D conversion I.



- Sampling is very slow.
- Accuracy is high: 20 – 24 bits.

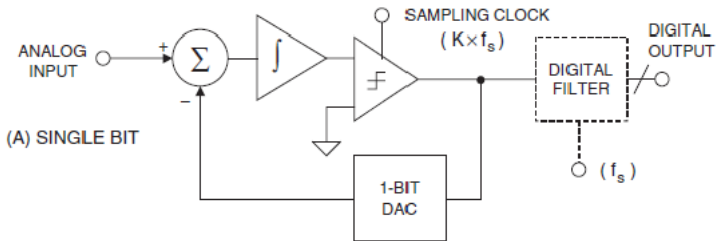
Dual-slope A/D conversion II.



- 1 The input **signal is connected to the input of the S/H**, the output of the integrator is set to zero.
- 2 The **conversion begins**: the signal is integrated for a length of N_{ref} clock cycles.
- 3 The **negative reference voltage is connected to the input** and the number of steps it takes (N_x) to discharge the capacitor is counted:

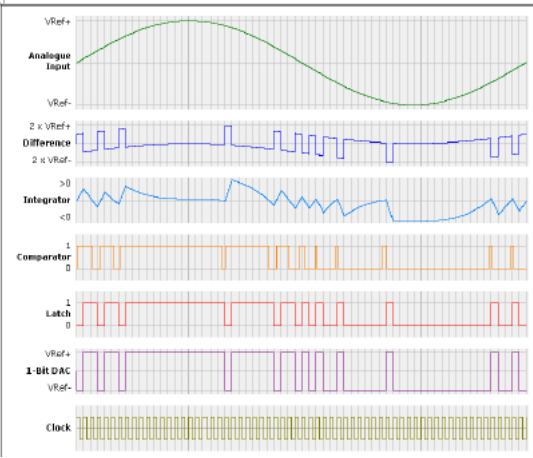
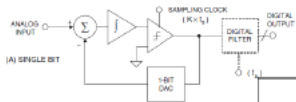
$$V_{in} = \frac{N_x}{N_{ref}} \cdot V_{ref}$$

Sigma-Delta ($\Sigma - \Delta$) A/D converters I.



- This is a first order $\Sigma - \Delta$ ADC.
- **Oversampling:** it samples at a much higher frequency than it is required by the Shannon-Nyquist theorem. The quantization noise is spread in a much larger frequency range this way.
- It is less sensitive to devices inaccuracies – easier to realize in an IC.
- An example: 24-bit ADC for sound input (0 – 20 kHz): 5th order, 64 \times oversampling.

Sigma-Delta ($\Sigma - \Delta$) A/D converters II.



Typical waveforms of a 1st order $\Sigma - \Delta$ ADC